12V～76V input voltage range 3A output current

1ch Buck Converter Integrated FET

BD9G341EFJ

General Description
The BD9G341EFJ is a buck switching regulator with integrated 150mΩ power MOSFET. Current mode architecture provides fast transient response and a simple phase compensation setup. The operating frequency is programmable from 50kHz to 750kHz. Additional protection features are included such as Over Current Protection, Thermal shutdown and Under voltage lockout. The under voltage lockout and hysteresis can be set by external resistor.

Features
- Wide input voltage range from 12V to 76V.
- Integrated 80V/3.5A/150mΩ NchFET.
- Current mode.
- Variable frequency from 50kHz to 750kHz.
- Accurate reference voltage. (1.0 V±1.5%).
- Precision EN threshold (+/-3%).
- Soft-start function
- 0uA Standby current
- Over Current Protection (OCP), Under Voltage Lockout(UVLO), Thermal-Shutdown(TSD), Over Voltage Protection (OVP)
- Thermally enhanced HTSOP-J8 package

Applications
- Industrial distributed power applications.
- Automotive Application
- Battery powered equipment.

Key specifications
- Input voltage: 12~76[V]
- Ref voltage(Ta=25°C) ±1.5[%] (Ta=-40~85°C) ±2.0[%]
- Max output current: 3 [A] (Max.)
- Operating Temperature: -40°C~85°C
- Max junction temperature: 150°C

Package(s)
HTSOP-J8  4.90mm x 6.00mm x 1.00mm

Datasheet
Figure 1. Typical Application Schematic
## Pin Description

<table>
<thead>
<tr>
<th>Pin No.</th>
<th>Pin Name</th>
<th>Description</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>LX</td>
<td>Switching node. It should be connected as near as possible to the schottky barrier diode, and inductor.</td>
</tr>
<tr>
<td>2</td>
<td>GND</td>
<td>Ground pin. GND pattern is kept from the current line of input capacitor to output capacitor.</td>
</tr>
<tr>
<td>3</td>
<td>VC</td>
<td>The output of the internal error amplifier. The phase compensation implementation is connected between this pin to GND.</td>
</tr>
<tr>
<td>4</td>
<td>FB</td>
<td>Voltage feedback pin. This pin is the error-amp input with the DC voltage is set at 0.75V with feed-back operation.</td>
</tr>
<tr>
<td>5</td>
<td>RT</td>
<td>The internal oscillator frequency set pin. The internal oscillator is set with a single resistor connected between this pin and the GND pin. Recommended frequency range is 50kHz to 750kHz</td>
</tr>
<tr>
<td>6</td>
<td>EN</td>
<td>Shutdown pin. If the voltage of this pin is below 0.8V, the regulator will be in a low power state. If the voltage of this pin is between 0.8V and 2.4V, the IC will be in standby mode. If the voltage of this pin is above 2.6V, the regulator is operational. An external voltage divider can be used to set under voltage threshold. If this pin is left open circuit when converter is operating. This pin output 10μA source current. If this pin is left open circuit, a 10μA pull up current source configures the regulator fully operational.</td>
</tr>
<tr>
<td>7</td>
<td>BST</td>
<td>Boost input for bootstrap capacitor. The external capacitor is required between the BST and the LX pin. A 0.1μF ceramic capacitor is recommended.</td>
</tr>
<tr>
<td>8</td>
<td>VCC</td>
<td>Input supply voltage pin.</td>
</tr>
<tr>
<td>-</td>
<td>Thermal Pad</td>
<td>Connect to GND.</td>
</tr>
</tbody>
</table>

Figure 2. Pin Configuration (TOP VIEW)
Description of Block(s)

1. **Reference**
   This block generates inner reference voltage.

2. **REG**
   This block generates 8V reference voltage for bootstrap.

3. **OSC**
   This block generates inner CLK.
   The internal oscillator is set with a single resistor connected between this pin and the GND pin.
   Recommended frequency range is 50 kHz to 750 kHz. If RT pin connect to 47kohm, frequency is set 200 kHz.

4. **Soft Start**
   Soft Start of the output voltage of regulator prevents in-rush current during Start-up.
   Soft Start time is 20msec (typ)

5. **ERROR AMP**
   This is an error amplifier what detects output signal, and outputs PWM control signal.
   Internal reference voltage is set to 1.0V.

6. **ICOMP**
   This is a comparator that outputs PWM signal from current feed-back signal and error-amp output for current-mode.

7. **Nch FET SW**
   This is a 80V/150mΩ-Power Nch MOSFET SW that converts inductor current of DC/DC converter
   Since the current rating of this FET is 3.5A, it should be used within 3.5A including the DC current and ripple current of the coil.

8. **UVLO**
   This is a Low Voltage Error Prevention Circuit.
   This prevents internal circuit error during increase of Power supply Voltage and during decline of Power supply Voltage.
   It monitors VCC Pin Voltage and internal REG Voltage. When VCC Voltage becomes 11V and below, UVLO turns OFF all Output FET and turns OFF the DC/DC Comparator Output, and the Soft Start Circuit resets.
   Now this Threshold has Hysteresis of 200mV.
9. EN
Shutdown function. If the voltage of this pin is below 0.8V, the regulator will be in a low power state. If the voltage of this pin is between 0.8V and 2.4V, it will be standby mode. If the voltage of this pin is above 2.6V, the regulator is operational. An external voltage divider can be used to set under voltage threshold. If this pin is left open circuit, when converter is operating. This pin output 10uA source current. If this pin is left open circuit, a 10uA pull up current source configures the regulator fully operational. When IC turn off, EN pin is pulled down by pull down resistor that sink above 10uA.

10. OCP
Over current protection
If the current of power MOSFET is over 6.0A (typ), this function reduces duty pulse by pulse and restricts the over current. If IC detects OCP 2 times sequentially, the device will stop and after 20 msec restart.

11. TSD
This is Thermal Shutdown Detection
When it detects an abnormal temperature exceeding Maximum Junction Temperature (Tj=150℃), it turns OFF all Output FETs, and turns OFF the DC/DC Comparator Output. When Temperature falls, and the IC automatically returns

12. OVP
Over voltage protection.
Output voltage is monitored with FB terminal, and output FET is turned off when it becomes 120% of set-point voltage.

### Absolute Maximum Ratings

<table>
<thead>
<tr>
<th>Item</th>
<th>Symbol</th>
<th>Ratings</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Maximum input voltage</td>
<td>VCC</td>
<td>80</td>
<td>V</td>
</tr>
<tr>
<td>BST</td>
<td>VBST</td>
<td>85</td>
<td>V</td>
</tr>
<tr>
<td>Maximum input current</td>
<td>Imax</td>
<td>3.5</td>
<td>A</td>
</tr>
<tr>
<td>BST to LX</td>
<td>VBST</td>
<td>15</td>
<td>V</td>
</tr>
<tr>
<td>EN</td>
<td>VEN</td>
<td>80</td>
<td>V</td>
</tr>
<tr>
<td>LX</td>
<td>VLX</td>
<td>80</td>
<td>V</td>
</tr>
<tr>
<td>FB</td>
<td>VFB</td>
<td>7</td>
<td>V</td>
</tr>
<tr>
<td>Power Dissipation</td>
<td>Pd</td>
<td>3.76</td>
<td>W</td>
</tr>
<tr>
<td>Operating Temperature</td>
<td>Topr</td>
<td>-40~+85</td>
<td>℃</td>
</tr>
<tr>
<td>Storage Temperature</td>
<td>Tstg</td>
<td>-55~+150</td>
<td>℃</td>
</tr>
<tr>
<td>Junction Temperature</td>
<td>Tjmax</td>
<td>150</td>
<td>℃</td>
</tr>
</tbody>
</table>

\(^{(NOTE1)}\)During mounting of 70×70×1.6t mm 4layer board. Reduce by 5.4mW for every 1℃ increase. (Above 25℃)
### Electrical Characteristics (Unless otherwise specified $T_a=25^\circ C$, $V_{CC}=48V$, $V_o=5V$, $EN=3V$, $RT=47\Omega$)

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Symbol</th>
<th>Limit</th>
<th>Unit</th>
<th>Condition</th>
</tr>
</thead>
<tbody>
<tr>
<td><strong>【Circuit Current】</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Stand-by current of VCC</td>
<td>$I_{st}$</td>
<td>—</td>
<td>0</td>
<td>10</td>
</tr>
<tr>
<td>Circuit current of VCC</td>
<td>$I_{cc}$</td>
<td>—</td>
<td>1.5</td>
<td>2.0</td>
</tr>
<tr>
<td><strong>【Under Voltage Lock Out (UVLO)】</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Detect Voltage</td>
<td>$V_{ccuv}$</td>
<td></td>
<td>10.4</td>
<td>11</td>
</tr>
<tr>
<td>Hysteresis width</td>
<td>$V_{uvhy}$</td>
<td>—</td>
<td>200</td>
<td>300</td>
</tr>
<tr>
<td><strong>【Error Amp】</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>FB threshold voltage</td>
<td>$V_{FBN}$</td>
<td></td>
<td>0.985</td>
<td>1.000</td>
</tr>
<tr>
<td></td>
<td></td>
<td></td>
<td>0.980</td>
<td>1.000</td>
</tr>
<tr>
<td>FB Input bias current</td>
<td>$I_{FB}$</td>
<td></td>
<td>-1</td>
<td>0</td>
</tr>
<tr>
<td>VC source current</td>
<td>$I_{source}$</td>
<td></td>
<td>15</td>
<td>40</td>
</tr>
<tr>
<td>VC sink current</td>
<td>$I_{sink}$</td>
<td></td>
<td>-65</td>
<td>-40</td>
</tr>
<tr>
<td>Soft start time</td>
<td>$T_{soft}$</td>
<td></td>
<td>15</td>
<td>20</td>
</tr>
<tr>
<td>Error amplifier DC gain</td>
<td>$AVEA$</td>
<td></td>
<td>—</td>
<td>10000</td>
</tr>
<tr>
<td>Trans conductance</td>
<td>$G_{EA}$</td>
<td></td>
<td>—</td>
<td>300</td>
</tr>
<tr>
<td><strong>【Current Sense Amp】</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>VC to switch current trans conductance</td>
<td>$G_{CS}$</td>
<td>—</td>
<td>10</td>
<td>—</td>
</tr>
<tr>
<td><strong>【OCP】</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Detect current</td>
<td>$I_{ocp}$</td>
<td></td>
<td>3.5</td>
<td>6.0</td>
</tr>
<tr>
<td>OCP latch count</td>
<td>$NOCP$</td>
<td>—</td>
<td>2</td>
<td>—</td>
</tr>
<tr>
<td>OCP latch hold time</td>
<td>$TOCP$</td>
<td></td>
<td>15</td>
<td>20</td>
</tr>
<tr>
<td><strong>【出力部】</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Lx NMOS ON resistance</td>
<td>$R_{onH}$</td>
<td></td>
<td>—</td>
<td>150</td>
</tr>
<tr>
<td><strong>【CTL】</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>EN Pin inner REG on voltage</td>
<td>$VENON$</td>
<td></td>
<td>1.3</td>
<td>—</td>
</tr>
<tr>
<td>EN Pin IC output on threshold</td>
<td>$VENUV$</td>
<td></td>
<td>2.52</td>
<td>2.6</td>
</tr>
<tr>
<td>EN pin</td>
<td>$I_{EN}$</td>
<td></td>
<td>9.0</td>
<td>10.0</td>
</tr>
<tr>
<td><strong>【Oscillator】</strong></td>
<td></td>
<td></td>
<td></td>
<td></td>
</tr>
<tr>
<td>Oscillator frequency</td>
<td>$F_{osc}$</td>
<td></td>
<td>180</td>
<td>200</td>
</tr>
<tr>
<td>Forced off time</td>
<td>$Toff$</td>
<td></td>
<td>—</td>
<td>—</td>
</tr>
</tbody>
</table>

### Recommended Operating Ratings ($T_a=25^\circ C$)

<table>
<thead>
<tr>
<th>Item</th>
<th>Symbol</th>
<th>Rating</th>
<th>Unit</th>
</tr>
</thead>
<tbody>
<tr>
<td>Power Supply Voltage</td>
<td>$V_{CC}$</td>
<td>12</td>
<td>—</td>
</tr>
<tr>
<td>Output voltage</td>
<td>$V_{OUT}$</td>
<td>$1.0^{(Note2)}$</td>
<td>—</td>
</tr>
<tr>
<td>Output current</td>
<td>$I_{OUT}$</td>
<td>—</td>
<td>—</td>
</tr>
<tr>
<td>Oscillator frequency</td>
<td>$F_{osc}$</td>
<td>50</td>
<td>—</td>
</tr>
</tbody>
</table>

*(Note2)* Restricted by $minduty=1\times MinOn\ Time (f:\ frequency)$

If the voltage of $Vcc \times minduty \leq 1V$, this value is minimum output.

*(Note3)* Restricted by $maxduty=1\times forced\ off\ time$

The maximum output is $Vcc \times maxduty - Iout \times R_{onH}$. 

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Typical Performance Characteristics
(Unless otherwise specified, Ta=25°C, VCC=24V, VOUT=5V)

Fig.4 Oscillator Frequency - Temperature
Fig.5 FB Threshold Voltage - Input Voltage
Fig.6 FB Threshold Voltage - Temperature
Fig.7 Forced off time - Temperature
Fig.8 UVLO Threshold Voltage - Temperature
Fig.9 OCP Detect Current - Temperature
Fig. 10 Soft Start Time - Temperature
Fig. 11 EN Pin Inner REG ON Threshold - Temperature
Fig. 12 ENUVLO Threshold - Temperature
Fig. 13 EN Source Current - Temperature
Fig. 14 NMOS ON Resistance - Temperature
Reference Characteristics of Typical Application Circuits

Vout=5V , f=200kHz

Parts:
L: SUMIDA CDRH129LD 33μH
C1: TDK C5750X7S2A106K 10μF/100V
C2: TDK C4532X5R0J107M 100μF/6.3V
D1: Rohm RB095B-90

Fig.15 Efficiency – Output Current
Fig. 16 Start-up Characteristics

Fig. 17 Load Response
I_out: 100mA ⇔ 1A

Fig. 18 Lx Switching/Vout Ripple
I_o = 100mA

Fig. 19 Lx Switching/Vout Ripple
I_o = 1A

Fig. 20 Frequency Response
I_o = 100mA

Fig. 21 Frequency Response
I_o = 3.0A
Reference Characteristics of Typical Application Circuits

Vout=3.3V, f=200kHz

Vin=12~76V

C1: 10uF/100V

R1 Ω

R2 Ω

L : 33µH

LX

C2: 100µF/6.3V

0.1µF

0.01µF

0.56kΩ

6.2kΩ

VCC

BST

EN

FB

VC

GND

RT

VOUT=3.3V /3A

Parts:

L : SUMIDA CDRH129LD 33µH

C1 : TDK C5750X7S2A106K 10µF/100V

C2 : TDK C4532X5R0J107M 100µF/6.3V

D1 : Rohm RB095B-90

Fig.22 Efficiency – Output Current
Fig.23 Start-up Characteristics

Fig.24 Load Response
Iout: 100mA ⇔ 1A

Fig.25 Lx Switching/Vout Ripple
Io = 100mA

Fig.26 Lx Switching/Vout Ripple
Io = 1A

Fig.27 Frequency Response
Io = 100mA

Fig.28 Frequency Response
Io = 3A
### Detailed Description

◇ Frequency setting

Arbitrary internal oscillator frequency setup is possible by connecting RT resistance. Recommended frequency range is 50 kHz to 750 kHz.

For setting frequency \( f \) [Hz], RT resistance \( RT \) is looked for using the following formula.

\[
RT = \frac{1 - 400 \times 10^{-9}}{f} \times 96.48 \times 10^{-12} \text{[\(\Omega\)]}
\]

If setting frequency is 200kHz, \( RT \) is 47kΩ.

RT resistance is related to frequency as shown in Figure 26.

![Oscillator Frequency - RT resistance](image)

◇ External UVLO threshold

The high precision reset function is built in EN terminal of BD9G341EFJ, and arbitrary low-voltage malfunction prevention setup is possible by connecting EN pin to resistance division of input voltage.

When you use, please set \( R1 \) and \( R2 \) to arbitrary UVLO threshold level \( V_{uv} \) and hysteresis \( V_{uvhys} \) like below.

\[
\begin{align*}
R1 &= \frac{V_{uvhys}}{I_{EN}} \text{[ohm]} \\
R2 &= \frac{V_{EN} \times R1}{V_{uv} - V_{VEN}} \text{[ohm]}
\end{align*}
\]

IEN:EN pin source current 10µA(typ)  VEN: EN pin output on threshold  2.6V(typ)

As an example in typical sample, When Vcc voltage which IC turned on 15V, Hysteresis width 1V, The resistance divider set to \( R1=100k\Omega \), \( R2=20k\Omega \).
OCP operation

The device has over current protection for protecting the FET from over current. To detect OCP 2 times sequentially, the device will stop and after 20msec restart.

VC

Lx

VOUT

OCP

OCP_LATCH

Fig.31 Timing chart at OCP operation
**Application Components Selection Method**

(1) Inductors

Something of the shield type that fulfills the current rating (Current value Ipeak below), with low DCR is recommended. Value of Inductance influences Inductor Ripple Current and becomes the cause of Output Ripple.

In the same way as the formula below, this Ripple Current can be made small for as big as the L value of Coil or as high as the Switching Frequency.

\[
I_{peak} = I_{OUT} + \frac{\Delta I_L}{2} \quad \cdots (1)
\]

\[
\Delta I_L = \frac{V_{CC} - V_{OUT}}{L} \times \frac{V_{OUT}}{V_{CC}} \times \frac{1}{f} \quad \cdots (2)
\]

(\(\Delta I_L\): Output Ripple Current, VCC: Input Voltage, VOUT: Output Voltage, f: Switching Frequency)

For design value of Inductor Ripple Current, please carry out design tentatively with about 20%~50% of Maximum Input Current.

In the BD9G341EFJ, it is recommended the below series of 4.7\(\mu\)H~33\(\mu\)H inductance value.

Recommended Inductor: SUMIDA CDRH127H Series

(2) Output Capacitor

In order for capacitor to be used in output to reduce output ripple, Low ceramic capacitor of ESR is recommended. Also, for capacitor rating, on top of putting into consideration DC Bias characteristics, please use something whose maximum rating has sufficient margin with respect to the Output Voltage.

Output ripple voltage is looked for using the following formula.

\[
V_{pp} = \Delta I_L \times \frac{1}{2\pi \times f \times C_{OUT}} + \Delta I_L \times R_{ESR} \quad \cdots (3)
\]

Please design in a way that it is held within Capacity Ripple Voltage.

In the BD9G341EFJ, it is recommended a ceramic capacitor over 10\(\mu\)F.

(3) Output voltage setting

The internal reference voltage of ERROR AMP is 1.0V. Output voltage is determined like (4) types.

\[
V_{OUT} = \frac{R_1 + R_2}{R_2} \quad \cdots (4)
\]

(4) Bootstrap Capacitor

Please connect from 0.1\(\mu\)F (Laminate Ceramic Capacitor) between BST Pin and Lx Pins.

(5) Catch Diode

BD9G341EFJ should be taken to connect external catch diode between Lx Pin and GND Pin. The diode require adherence to absolute maximum Ratings of application. Opposite direction voltage should be higher than maximum voltage of Lx Pin (VCCMAX + 0.5V). The peak current is required to be higher than IOUTMAX + \(\Delta I_L\).

(6) Input Capacitor

BD9G341EFJ needs an input decoupling capacitor. It is recommended a low ceramic capacitor ESR over 4.7\(\mu\)F. Additionally, it should be located as close as possible.

Capacitor should be selected by maximum input voltage with input ripple voltage.

Input ripple voltage is calculated by using the following formula.

\[
\Delta V_{CC} = \frac{I_{OUT}}{f \times C_{VCC}} \times \frac{V_{OUT}}{V_{CC}} \times \left[ 1 - \frac{V_{OUT}}{V_{CC}} \right] \quad \cdots (5)
\]
CVCC: Input capacitor
RMS ripple current is calculated by using the following formula.

\[ I_{CVCC} = I_{OUT} \times \sqrt{\frac{V_{OUT}}{VCC} \times (1 - \frac{V_{OUT}}{VCC})} \]  \( \cdots (6) \)

If \( VCC = 2 \times VOUT \), RMS ripple current is maximum. That is determined by \( (9) \).

\[ I_{CVCC,\text{max}} = \frac{I_{OUT}}{2} \]  \( \cdots (7) \)

(7) About Adjustment of DC/DC Comparator Frequency Characteristics
Role of Phase compensation element C1, C2, R3

Stability and Responsiveness of Loop are controlled through VC Pin which is the output of Error Amp.
The combination of zero and pole that determines Stability and Responsiveness is adjusted by the combination of resistor and capacitor that are connected in series to the VC Pin.

DC Gain of Voltage Return Loop can be calculated for using the following formula.

\[ Adc = Rl \times G_{CS} \times A_{VEA} \times \frac{VFB}{VOUT} \]  \( \cdots (8) \)

Here, \( VFB \) is Feedback Voltage (1.0V), \( A_{EA} \) is Voltage Gain of Error amplifier (typ: 55.6dB),
\( G_{cs} \) is the Trans-conductance of Current Detect (typ: 10A/V), and \( Rl \) is the Output Load Resistance value.

There are 2 important poles in the Control Loop of this DC/DC.
The first occurs with/ through the output resistance of Phase compensation Capacitor (C1) and Error amplifier.
The other one occurs with/through the Output Capacitor and Load Resistor.
These poles appear in the frequency written below.

\[ fp1 = \frac{G_{EA}}{2\pi \times C1 \times A_{VEA}} \]  \( \cdots (9) \)

\[ fp2 = \frac{1}{2\pi \times COUT \times Rl} \]  \( \cdots (10) \)

Here, \( G_{EA} \) is the trans-conductance of Error amplifier (typ: 300 \( \mu \)A/V).

Here, in this Control Loop, one zero becomes important. With the zero which occurs because of Phase compensation Capacitor.
C1 and Phase compensation Resistor R3, the Frequency below appears.

\[ f_{z1} = \frac{1}{2\pi \times C1 \times R3} \quad \cdots (11) \]

Also, if Output Capacitor is big, and that ESR (RESR) is big, in this Control Loop, there are cases when it has an important, separate zero (ESR zero).

This ESR zero occurs due to ESR of Output Capacitor and Capacitance, and exists in the Frequency below.

\[ f_{z_{ESR}} = \frac{1}{2\pi \times COUT \times RESR} \quad \cdots (12) \quad \text{(ESR zero)} \]

In this case, the 3\textsuperscript{rd} pole determined with the 2\textsuperscript{nd} Phase compensation Capacitor (C2) and Phase Correction Resistor (R3) is used in order to correct the ESR zero results in Loop Gain.

This pole exists in the frequency shown below.

\[ fp3 = \frac{1}{2\pi \times C2 \times R3} \quad \cdots (13) \quad \text{(Pole that corrects ESR zero)} \]

The target of Phase compensation design is to create a communication function in order to acquire necessary band and Phase margin.

Cross-over Frequency (band) at which Loop gain of Return Loop becomes "0" is important. When Cross-over Frequency becomes low, Power supply fluctuation response, Load response, etc. worsens. On the other hand, when Cross-over Frequency is too high, instability of the Loop can occur. Tentatively, Cross-over Frequency is targeted to be made 1/20 or below of Switching Frequency.

Selection method of Phase Compensation constant is shown below.

1. Phase Compensation Resistor (R3) is selected in order to set to the desired Cross-over Frequency. Calculation of RC is done using the formula below.

\[ R3 = \frac{2\pi \times COUT \times fc}{G_{EA} \times G_{CS}} \times \frac{VOUT}{VFB} \quad \cdots (14) \]

Here, fc is the desired Cross-over Frequency. It is made about 1/20 and below of the Normal Switching Frequency (fs).

2. Phase compensation Capacitor (C1) is selected in order to achieve the desired phase margin.

In an application that has a representative Inductance value (about several 3.3 \( \mu \)H ~ 10 \( \mu \)H), by matching zero of compensation to 1/4 and below of the Cross-over Frequency, sufficient Phase margin can be acquired. C1 can be calculated using the following formula.

\[ C1 > \frac{4}{2\pi \times R3 \times fc} \quad \cdots (15) \]

3. Examination whether the second Phase compensation Capacitor C2 is necessary or not is done.

If the ESR zero of Output Capacitor exists in a place that is smaller than half of the Switching Frequency, a second Phase compensation Capacitor is necessary. In other words, it is the case wherein the formula below happens.

\[ \frac{1}{2\pi \times COUT \times RESR} < \frac{fs}{2} \quad \cdots (16) \]

In this case, add the second Phase compensation Capacitor C2, and match the frequency of the third pole to the Frequency fp3 of ESR zero.

\[ C2 = \frac{COUT \times RESR}{R3} \quad \cdots (17) \]
**PCB Layout**

Layout is a critical portion of good power supply design. There are several signals paths that conduct fast changing currents or voltages that can interact with stray inductance or parasitic capacitance to generate noise or degrade the power supplies performance. To help eliminate these problems, the VCC pin should be bypassed to ground with a low ESR ceramic bypass capacitor with B dielectric. Care should be taken to minimize the loop area formed by the bypass capacitor connections, the VCC pin, and the anode of the catch diode. See Fig.28 for a PCB layout example. The GND pin should be tied directly to the thermal pad under the IC and the thermal pad. In order to reduce the influence of the impedance and L of the parasitic, the high current line is thick and short.

Input decoupling capacitor should be located as close to the VCC pins.

In order to minimize the parasitic capacitor and impedance of pattern, catch diode and inductance should be located as close to the Lx pin.

The thermal pad should be connected to any internal PCB ground planes using multiple VIAs directly under the IC.

GND feedback resistor, phase compensation element and RT resistor don’t give the common impedance resistor against high current line.

---

**Figure 35. Evaluation Board Pattern**
Power Dissipation

It is shown below reducing characteristics of power dissipation to mount 70mm×70mm×1.6mm³ PCB. Junction temperature must be designed not to exceed 150°C.

Power Dissipation Estimate

The following formulas show how to estimate the device power dissipation under continuous mode operations. They should not be used if the device is working in the discontinuous conduction mode.

The device power dissipation includes:

1) Conduction loss: \( P_{\text{con}} = I_{\text{OUT}}^2 \times R_{\text{onH}} \times V_{\text{OUT}} / V_{\text{CC}} \)
2) Switching loss: \( P_{\text{sw}} = 16n \times V_{\text{CC}} \times I_{\text{OUT}} \times f_{\text{sw}} \)
3) Gate charge loss: \( P_{\text{gc}} = 500p \times 7 \times f_{\text{sw}} \)
4) Quiescent current loss: \( P_{\text{q}} = 1.5m \times V_{\text{CC}} \)

Where:

- \( I_{\text{OUT}} \) is the output current (A), \( R_{\text{onH}} \) is the on-resistance of the high-side MOSFET (Ω), \( V_{\text{OUT}} \) is the output voltage (V), \( V_{\text{CC}} \) is the input voltage (V), \( f_{\text{sw}} \) is the switching frequency (Hz).

Therefore

Power dissipation of IC is the sum of above dissipation.
\( P_{\text{d}} = P_{\text{con}} + P_{\text{sw}} + P_{\text{gc}} + P_{\text{q}} \)

For given \( T_{\text{j}} \), \( T_{\text{j}} = T_{\text{a}} + 8j_{\text{a}} \times P_{\text{d}} \)

Where:

- \( P_{\text{d}} \) is the total device power dissipation (W), \( T_{\text{a}} \) is the ambient temperature (°C)
- \( T_{\text{j}} \) is the junction temperature (°C), \( 8j_{\text{a}} \) is the thermal resistance of the package (°C)
## I/O Equivalent Schematic

<table>
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<tr>
<th>Pin. No</th>
<th>端子名</th>
<th>端子等価回路図</th>
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<tr>
<td>1</td>
<td>Lx</td>
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<td>2</td>
<td>GND</td>
<td><img src="#" alt="VCC" /></td>
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<tr>
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<td>BST</td>
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<td>8</td>
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<tr>
<td>6</td>
<td>EN</td>
<td><img src="#" alt="EN" /></td>
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</tbody>
</table>

![RT](#)

![VC](#)

![FB](#)

![VCC](#)

![GND](#)
Operational Notes

1. **Reverse Connection of Power Supply**
   Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply, such as mounting an external diode between the power supply and the IC’s power supply pins.

2. **Power Supply Lines**
   Design the PCB layout pattern to provide low impedance supply lines. Separate the ground and supply lines of the digital and analog blocks to prevent noise in the ground and supply lines of the digital block from affecting the analog block. Furthermore, connect a capacitor to ground at all power supply pins. Consider the effect of temperature and aging on the capacitance value when using electrolytic capacitors.

3. **Ground Voltage**
   Ensure that no pins are at a voltage below that of the ground pin at any time, even during transient condition. OR

4. **Ground Wiring Pattern**
   When using both small-signal and large-current ground traces, the two ground traces should be routed separately but connected to a single ground at the reference point of the application board to avoid fluctuations in the small-signal ground caused by large currents. Also ensure that the ground traces of external components do not cause variations on the ground voltage. The ground lines must be as short and thick as possible to reduce line impedance.

5. **Thermal Consideration**
   Should by any chance the power dissipation rating be exceeded the rise in temperature of the chip may result in deterioration of the properties of the chip. The absolute maximum rating of the Pd stated in this specification is when the IC is mounted on a 70mm x 70mm x 1.6mm glass epoxy board. In case of exceeding this absolute maximum rating, increase the board size and copper area to prevent exceeding the Pd rating.

6. **Recommended Operating Conditions**
   These conditions represent a range within which the expected characteristics of the IC can be approximately obtained. The electrical characteristics are guaranteed under the conditions of each parameter.

7. **Inrush Current**
   When power is first supplied to the IC, it is possible that the internal logic may be unstable and inrush current may flow instantaneously due to the internal powering sequence and delays, especially if the IC has more than one power supply. Therefore, give special consideration to power coupling capacitance, power wiring, width of ground wiring, and routing of connections.

8. **Operation Under Strong Electromagnetic Field**
   Operating the IC in the presence of a strong electromagnetic field may cause the IC to malfunction.

9. **Testing on Application Boards**
   When testing the IC on an application board, connecting a capacitor directly to a low-impedance output pin may subject the IC to stress. Always discharge capacitors completely after each process or step. The IC’s power supply should always be turned off completely before connecting or removing it from the test setup during the inspection process. To prevent damage from static discharge, ground the IC during assembly and use similar precautions during transport and storage.

10. **Inter-pin Short and Mounting Errors**
    Ensure that the direction and position are correct when mounting the IC on the PCB. Incorrect mounting may result in damaging the IC. Avoid nearby pins being shorted to each other especially to ground, power supply and output pin. Inter-pin shorts could be due to many reasons such as metal particles, water droplets (in very humid environment) and unintentional solder bridge deposited in between pins during assembly to name a few.
11. Unused Input Pins
Input pins of an IC are often connected to the gate of a MOS transistor. The gate has extremely high impedance and extremely low capacitance. If left unconnected, the electric field from the outside can easily charge it. The small charge acquired in this way is enough to produce a significant effect on the conduction through the transistor and cause unexpected operation of the IC. So unless otherwise specified, unused input pins should be connected to the power supply or ground line.

12. Regarding the Input Pin of the IC
This monolithic IC contains P+ isolation and P substrate layers between adjacent elements in order to keep them isolated. P-N junctions are formed at the intersection of the P layers with the N layers of other elements, creating a parasitic diode or transistor. For example (refer to figure below):

When \( \text{GND} > \text{Pin A} \) and \( \text{GND} > \text{Pin B} \), the P-N junction operates as a parasitic diode.
When \( \text{GND} > \text{Pin B} \), the P-N junction operates as a parasitic transistor.

Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, operational faults, or physical damage. Therefore, conditions that cause these diodes to operate, such as applying a voltage lower than the GND voltage to an input pin (and thus to the P substrate) should be avoided.

![Figure 37. Example of monolithic IC structure](image)

13. Ceramic Capacitor
When using a ceramic capacitor, determine the dielectric constant considering the change of capacitance with temperature and the decrease in nominal capacitance due to DC bias and others.

14. Area of Safe Operation (ASO)
Operate the IC such that the output voltage, output current, and power dissipation are all within the Area of Safe Operation (ASO).

Operational Notes – continued

15. Thermal Shutdown Circuit (TSD)
This IC has a built-in thermal shutdown circuit that prevents heat damage to the IC. Normal operation should always be within the IC's power dissipation rating. If however the rating is exceeded for a continued period, the junction temperature \( (T_j) \) will rise which will activate the TSD circuit that will turn OFF all output pins. When the \( T_j \) falls below the TSD threshold, the circuits are automatically restored to normal operation.

Note that the TSD circuit operates in a situation that exceeds the absolute maximum ratings and therefore, under no circumstances, should the TSD circuit be used in a set design or for any purpose other than protecting the IC from heat damage.

16. Over Current Protection Circuit (OCP)
This IC incorporates an integrated overcurrent protection circuit that is activated when the load is shorted. This protection circuit is effective in preventing damage due to sudden and unexpected incidents. However, the IC should not be used in applications characterized by continuous operation or transitioning of the protection circuit.
**Ordering Information**

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<th>1</th>
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<td>Embossed tape and reel</td>
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**Marking Diagrams**

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<tr>
<td>4.90mm x 6.00mm x 1.00mm</td>
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**HTSOP-J8**

- (TOP VIEW)
- Part Number Marking
- LOT Number
- 1PIN MARK
Physical Dimension, Tape and Reel Information

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<thead>
<tr>
<th>Package Name</th>
<th>HTSOP-J8</th>
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</table>

![Diagram of HTSOP-J8 package](image)

**Physical Dimension**

- **1 PIN MARK**
- **0.545**
- **3.9±0.1**
- **4.9±0.2**
- **6.5±0.2**
- **0.42±0.05**
- **0.08±0.05**
- **1.05±0.15**
- **0.17±0.05**
- **0.08±0.03**
- **4°±6°**

**Tape and Reel Information**

- **PKG:** HTSOP-J8
- **Drawing No.:** EX169-5002-2

**Packaging Information**

- **Type:**エンボステーピング
- **Quantity:** 2500 pcs
- **Handling:**
  - Reiを左手に持ち、右手でテープを引き出したときに（製品の1番ビンが左上になる方向）
### Revision History

<table>
<thead>
<tr>
<th>Date</th>
<th>Revision</th>
<th>Changes</th>
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<tbody>
<tr>
<td>15 Jul 2014</td>
<td>001</td>
<td>New Release</td>
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**Revision History**

- **Date**: 15 Jul 2014
- **Revision**: 001
- **Changes**: New Release
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   [b] Installation of redundant circuits to reduce the impact of single or multiple circuit failure

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   [b] Use of our Products outdoors or in places where the Products are exposed to direct sunlight or dust
   [c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl₂, H₂S, NH₃, SO₂, and NO₂
   [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
   [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
   [f] Sealing or coating our Products with resin or other coating materials
   [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
   [h] Use of the Products in places subject to dew condensation

4. The Products are not subject to radiation-proof design.

5. Please verify and confirm characteristics of the final or mounted products in using the Products.

6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.

7. De-rate Power Dissipation (Pd) depending on Ambient temperature (Ta). When used in sealed area, confirm the actual ambient temperature.

8. Confirm that operation temperature is within the specified range described in the product specification.

9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

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2. In principle, the reflow soldering method must be used; if flow soldering method is preferred, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification
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   - the temperature or humidity exceeds those recommended by ROHM
   - the Products are exposed to direct sunshine or condensation
   - the Products are exposed to high Electrostatic

2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.

3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.

4. Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

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